# Alias Codes for Sparse Fourier Transforms 

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Joint Work with Sameer Pawar and Xiao (Simon) Li
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- Quentin Byron
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MRI


Images


Cognitive Radio


Speech \& audio


Astronomy


Radar


Ultrasound


GPS

## Outline

- Part I: Noiseless Recovery
- Sparse-Graph Alias Codes
- Part II: Noisy Recovery
- Sample-Optimal Recovery with Near Linear Run-time
- Near Sample-Optimal Recovery with Sub-linear Run-time


## PART I: <br> Noiseless Recovery

## Computing Sparse DFT | Problem Formulation

- Compute the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$ with $K \ll N$ :

$$
\begin{aligned}
x[n] & =\frac{1}{\sqrt{N}} \sum_{k \in \mathcal{K}} X[k] e^{\mathrm{i} \frac{2 \pi k}{N} n} \quad n=0, \cdots, N-1 \\
\mathcal{K} & =\text { chosen from }[N] \text { uniformly at random }
\end{aligned}
$$

- Classical solution: FFT algorithm
- Sample cost $=N$
- Computational cost $=\mathcal{O}(N \log N)$
- With sparsity, what are the fundamental bounds for
- Sample cost?
- Computational cost?


## Computing Sparse DFT | Numerical Phantoms for Cardiovascular MR



E—【 http://www.biomed.ee.ethz.ch/research/bioimaging/cardiac/mrxcat
Eidgenössische Technische Hochschule Zürich
Swiss Federal Institute of Technology Zurich

$$
\begin{aligned}
& 336=16 \times 21 \\
& 323=17 \times 19
\end{aligned}
$$


temporal difference across different frames of the phantom

## Computing Sparse DFT | Numerical Phantoms for Cardiovascular MR



## Computing Sparse DFT | Problem Formulation

- Compute the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$ with $K \ll N$ :

$$
\begin{aligned}
x[n] & =\frac{1}{\sqrt{N}} \sum_{k \in \mathcal{K}} X[k] e^{\mathrm{i} \frac{2 \pi k}{N} n}+w[n], \quad n=0, \cdots, N-1 \\
\mathcal{K} & =\text { chosen from }[N] \text { uniformly at random }
\end{aligned}
$$

## Assumptions and caveats:

- $X[k]$ is from a finite constellation.
- Sparsity $K=|\mathcal{K}|=\mathcal{O}\left(N^{\delta}\right)$ is sub-linear for some $\delta \in(0,1)$.
- Noise $w[n]$ is independent complex Gaussian $\mathcal{C N}\left(0, \sigma^{2}\right)$
- Recovery guarantees are probabilistic.


## Computing Sparse DFT | Problem Formulation

- Compute the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$ with $K \ll N$ :

$$
x[n]=\frac{1}{\sqrt{N}} \sum_{k \in \mathcal{K}} X[k] e^{\mathrm{i} \frac{2 \pi k}{N} n}+w[n], \quad n=0, \cdots, N-1
$$

$$
\mathcal{K}=\text { chosen from }[N] \text { uniformly at random }
$$

Theorem: FFAST algorithm computes the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$,

- using $M=O(K)$ samples,
- in $O(K \log K)$ computations,
- with probability at least $1-O(1 / M)$.


## Computing Sparse DFT \| Related Worlk

## Statistical Signal Processing

- Frequency estimation
- Prony [1795]
- Pisarenko ['73]
- Vetterli et al. ['02]
- more ...
- Subspace methods ['86]
- MUSIC, ESPRIT etc.



## Compressed Sensing

- Feng and Bresler [1996]
- Candes et. al [2006]
- Rauhut [2008]
- Wainwright [2009]
- Tang et. al [2012]
- more ...



## Sparse DFT

- Gilbert et. al [2002-2008]
- Mishali and Eldar [2010]
- Iwen [2010]
- Indyk et al. [2012]
- more ...



## Computing Sparse DFT \| Related Work

- More recent work on computing sparse DFT
- http://groups.csail.mit.edu/netmit/sFFT/paper.html
- Recent advances in compressed sensing and sketching methods
- Our method:
- targets support recovery instead of $\ell_{2} / \ell_{1}$ approximation
- uses the design and analysis of sparse-graph codes
- leverages harmonic retrieval methods in statistical signal processing


## Computing Sparse DFT | Insights

- Computes an exactly $K$-sparse $N$-length DFT using $\mathcal{O}(K)$ samples with $\mathcal{O}(K \log K)$ computations.
- A common framework for noiseless and noisy observations.



## Computing Sparse DFT | Insights

sub-sampling
below Nyquist rate
clever sub-sampling (for sparse case)

Chinese-Remainder-Theorem guided subsampling

 guided subsampling


Aliasing in the frequency domain
good "alias code"?

Sparse graph codes

## Computing Sparse DFT | <br> Insights

sub-sampling
below Nyquist rate
clever sub-sampling (for sparse case)

Chinese-Remainder-Theorem guided subsampling

good "alias code"?

Sparse graph codes

Coding-theoretic tools
> Design:

- Randomized constructions of good sparse-graph codes.
> Analysis:
- Density evolution
- Martingale

E Expander graph theory

## Computing Sparse DFT | Main Idea

time-domain $x[n]$ length $N=20$
frequency-domain $X[k]$ sparsity $K=5$




## Computing Sparse DFT | Main Idea

time-domain $x[n]$ length $N=20$
frequency-domain $X[k]$ sparsity $K=5$


## Computing Sparse DFT | Main Idea

```
time-domain }x[n] length N=2
```

frequency-domain $X[k]$ sparsity $K=5$


## Computing Sparse DFT | Main Idea

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## Computing Sparse DFT | Main Idea

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time-domain \(x[n]\) length \(N=20\)
```

frequency-domain $X[k]$ sparsity $K=5$



```
\downarrow 5
```

subsample by 5



## Computing Sparse DFT | Main Idea

time-domain $x[n]$ length $N=20$
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## Computing Sparse DFT | Main Idea

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## Computing Sparse DFT | Main Idea

time-domain $x[n]$ length $N=20$
frequency-domain $X[k]$ sparsity $K=5$


```
\downarrow
```

subsample by 5




## Computing Sparse DFT | Main Idea

```
time-domain \(x[n]\) length \(N=20\)
```

frequency-domain $X[k]$ sparsity $K=5$


$X[1]=1$

## Computing Sparse DFT | Main Idea

```
time-domain x[n] length N=20
```

frequency-domain $X[k]$ sparsity $K=5$

$\downarrow 5$
subsample by 5

$=$ DFT


## Computing Sparse DFT \| Main Idea

```
time-domain }x[n] length N=2
```

frequency-domain $X[k]$ sparsity $K=5$


```
\downarrow 5
```

shift \& subsample by 5


## Computing Sparse DFT | Main Idea

```
time-domain }x[n] length N=2
```

frequency-domain $X[k]$ sparsity $K=5$

$X[1] \quad X[3] \quad X[5] \quad X[10] \quad X[13]$
shift \& subsample by 5


## Our Measurements

$$
U_{S}[0] \quad U_{S}[1] \quad U_{S}[2] \quad U_{S}[3]
$$

## Computing Sparse DFT | Main Idea

```
time-domain }x[n] length N=2
```

frequency-domain $X[k]$ sparsity $K=5$


```
\downarrow 5
```

shift \& subsample by 5


zero-ton multi-ton single-ton single-ton

## Computing Sparse DFT | Main Idea

 subsample by 5

zero-ton multi-ton single-ton single-ton

$$
\downarrow 5
$$

shift \& subsample by 5


zero-ton multi-ton single-ton single-ton

zero-ton multi-ton single-ton single-ton


Kay, Steven. "A fast and accurate single frequency estimator." Acoustics, Speech and Signal Processing, IEEE Transactions on 37.12 (1989): 1987-1990.




Computing Sparse DFT | Main Idea


Computing Sparse DFT | Main Idea


Computing Sparse DFT | Main Idea





## Computing Sparse DFT | Main Idea

Stage 1
downsample by 5


peeling decoder

## Computing Sparse DFT | Main Idea

Stage 1 downsample by 5


Stage 2
downsample by 4


## Sparse DFT Computation = Decoding over Sparse Graphs



- Explicit graph: design well-understood.
- $(N-K)$ correctly received packets.
- $K$ erased packets.
- Peeling decoder recovers values.

non-zero DFT Coefficients
aliased frequency bins
- Implicit graph induced by careful sub-sampling
- $(N-K)$ zero DFT coefficients.
- $K$ unknown non-zero DFT coefficients.
- Peeling decoder recovers values \& locations.


## CRT-guided Subsampling Induces Good Graphs


> Chinese-Remainder-Theorem:
A number between 0-19 is uniquely represented by its remainders modulo $(4,5)$.
> Two graph ensembles are equivalent.

## More on Subsampling | Choosing Sulb-sampling Patterns

## Chinese Remainder Theorem

## $49 \times 50 \times 51$

Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2:
- Stage 3:



## More on Subsampling | Choosing Sulb-sampling Patterns

## Chinese Remainder Theorem

## $49 \times 50 \times 51$

Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3:



## 51

## More on Subsampling | Choosing Sulb-sampling Patterns

## Chinese Remainder Theorem

## $49 \times 50 \times 51$

Signal sparsity $K=N^{\delta}$ with $N=124950$

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- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51



## 51

## More on Subsampling | Choosing Sulb-sampling Patterns

Chinese Remainder Theorem

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Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51
- $\delta=2 / 3$ such that $K \approx 2500$



## More on Subsampling | Choosing Sulb-sampling Patterns

Chinese Remainder Theorem

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Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51
- $\delta=2 / 3$ such that $K \approx 2500$



## More on Subsampling | Choosing Sulb-sampling Patterns

Chinese Remainder Theorem

## $49 \times 50 \times 51$

Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51
- $\delta=2 / 3$ such that $K \approx 2500$
- Stage 1: subsample by 49 , keep $50 \times 51$
- Stage 2:
- Stage 3:


$$
51
$$

## More on Subsampling | Choosing Sulb-sampling Patterns

Chinese Remainder Theorem

## $49 \times 50 \times 51$

Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51
- $\delta=2 / 3$ such that $K \approx 2500$
- Stage 1: subsample by 49 , keep $50 \times 51$
- Stage 2: subsample by 50 , keep $49 \times 51$
- Stage 3:



## 51

## More on Subsampling | Choosing Sulb-sampling Patterns

Chinese Remainder Theorem

$$
49 \times 50 \times 51
$$

Signal sparsity $K=N^{\delta}$ with $N=124950$

- $\delta=1 / 3$ such that $K \approx 50$
- Stage 1: subsample by $50 \times 51$, keep 49
- Stage 2: subsample by $49 \times 51$, keep 50
- Stage 3: subsample by $49 \times 50$, keep 51


## 50

## This can be generalized to any sparsity index between 0 and 1

- Stage 3: subsample by 51, keep $49 \times 50$


## Algorithm Analysis | A Hitchhilker's Guide



Goal: prove that the algorithm finishes $K d$ steps
$K d$ edges to be removed


## Algorithm Analysis | A Hitchhilker's Guide



- Pick an arbitrary edge in the graph $(c, v)$.


## Algorithm Analysis | A Hitchhilker's Guide



- Examine its directed neighborhood at depth- $2 \ell$


## Algorithm Analysis | A Hitchhilker's Guide



- Examine its directed neighborhood at depth-2l


## Density Evolution | A Hitchhilker's Guide


$p_{\ell}=$ probability of being present at depth- $2 \ell$

Density Evolution | A Hitchhilker's Guide


Density Evolution | A Hitchhilker's Guide


Density Evolution | A Hitchhilker's Guide


Density Evolution | A Hitchhilker's Guide


$$
p_{\ell}=\left[1-\left(1-p_{\ell-1}\right)^{3}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right]
$$

## Density Evolution | A Hitchhilker's Guide



$$
p_{\ell}=\left[1-\left(1-p_{\ell-1}\right)^{3}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right]
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- It generalizes to $d$ stages:


## Density Evolution | A Hitchhilker's Guide



$$
p_{\ell}=\left[1-\left(1-p_{\ell-1}\right)^{3}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right] \times\left[1-\left(1-p_{\ell-1}\right)^{2}\right]
$$

- It generalizes to $d$ stages:

$$
p_{\ell}=\left(1-e^{-\frac{2 K d}{M} p_{\ell-1}}\right)^{d-1}
$$

- $K=$ sparsity
- $M=\#$ of samples
- $d=\#$ of stages


## Density Evolution | A Hitchhilker's Guide

| $d$ | 2 | 3 | 4 | 5 | 6 |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $M / 2 K$ | 2.0000 | 1.2219 | 1.2948 | 1.4250 | 1.5696 |



$$
p_{\ell}=\left(1-e^{-\frac{2 K d}{M} p_{\ell-1}}\right)^{d-1}
$$

- $K=$ sparsity
- $M=\#$ of samples
- $d=\#$ of stages


## Sampling Rate | Noiseless Setting: Theory versus Practice



- $N=7.7$ million
- $K=400$
- $d=3$ stages
- $M=1248$ samples


## Density Evolution

$$
p_{\ell}=\left(1-e^{-\frac{2 K d}{M} p_{\ell-1}}\right)^{d-1}
$$

## Algorithm Analysis | A Hitchhilker's Guide

## - Density Evolution

- assumes that the directed neighborhood is a tree
- tree-based average analysis

$$
p_{\ell}=\left(1-e^{-\frac{2 d K}{M} p_{\ell-1}}\right)^{d-1}
$$

$p_{\ell}$ can be made arbitrarily small

## Algorithm Analysis | A Hitchhilker's Guide

## - Density Evolution

- assumes that the directed neighborhood is a tree
- tree-based average analysis
$p_{\ell}=\left(1-e^{-\frac{2 d K}{M} p_{\ell-1}}\right)^{d-1}$
$p_{\ell}$ can be made arbitrarily small



## Algorithm Analysis | A Hitchhilker's Guide

## - Performance Concentration

- overall average analysis $p_{\ell}^{\star}$ without tree assumption

$$
\left|p_{\ell}^{\star}-p_{\ell}\right|<\epsilon_{1}, \forall \epsilon_{1}>0
$$

- actual performance concentrates around overall average analysis $\mathbb{P}\left(\mid \#\right.$ of actual remaining edges $\left.-K d p_{\ell}^{\star} \mid>\epsilon_{2}\right) \rightarrow 0, \quad \forall \epsilon_{2}>0$

$K d\left(1-p_{\ell}\right)$ edges removed



## Algorithm Analysis | A Hitchhilker's Guide



## Algorithm Analysis | A Hitchhilker's Guide

## - Expander Graph

- the remaining $K d p_{\ell}$ edges form an expander graph
- expander graphs guarantee steady supplies of single-tons
success with high probability!



## Peeling Performance |Numerical Examples


$600 \times 493$

## Peeling Performance |Numerical Examples



- $N=600 \times 493 \approx 300$ (thousand)
- $K \approx 25$ (thousand)
- $M \approx 65$ (thousand)

remaining image





remaining image



remaining image



remaining image



remaining image



remaining image



remaining image



remaining image




## PART II: <br> Noisy Recovery

Noisy Setting: R-FFAST | Signall Model



- $y[n]=x[n]+w[n]$, where $w[n] \in \mathcal{C N}\left(0, \sigma^{2}\right)$.
- There are $K$ non-zero $X[k]$, where $X[k]$ is from a finite constellation
- SNR is $\mathbb{E}|x[n]|^{2} / \mathbb{E}|w[n]|^{2}$ (e.g., $\mathrm{SNR}=0 \mathrm{~dB}$ )

Noisy Setting: R-FFAST | From Noiseless to Noisy Noiseless - FFAST


Noisy Setting: R-FFAST \| From Noiseless to Noisy

Noiseless - FFAST


Noisy - R-FFAST


## Noisy Setting: R-FFAST | From Noiseless to Noisy <br> Noisy - R-FFAST

Two schemes to choose shifts:


## Noisy Setting: R-FFAST | From Noiseless to Noisy <br> Noisy - R-FFAST

Two schemes to choose shifts:

- scheme 1:
- sample-optimal recovery $M=O(K \log N)$
- near-linear run-time $T=O(N \log N)$



## Noisy Setting: R-FFAST | From Noiseless to Noisy

Noisy - R-FFAST
Two schemes to choose shifts:

- scheme 1 :
- sample-optimal recovery $M=O(K \log N)$
- near-linear run-time

$$
T=O(N \log N)
$$

- scheme 2 :
- near sample-optimal recovery

$$
M=O\left(K \log ^{1 . \dot{3}} N\right)
$$

- sub-linear run-time

$$
M=O\left(K \log ^{2 . \dot{3}} N\right)
$$



## Architecture | Recap on FFAST Sampling




each DFT coefficient
represents a distinct frequency

## Architecture | Recap on FFAST Sampling




- The key is to pinpoint single-tons in terms of
- location
- value
- Equivalent to the parameters of a discrete sinusoid
- frequency
- amplitude
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \cdots & W^{19 \times 19}\end{array}\right]$
$N \times N \mathrm{DFT}$ matrix
$W=e^{-\mathrm{i} \frac{2 \pi}{N}}$ with $N=20$


## Architecture | From Noiseless to Noisy



Noiseless


Noisy
matching 2 points of a noiseless sinusoid
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ \hline 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \cdots & W^{19 \times 19}\end{array}\right]$

## Architecture | From Noiseless to Noisy



Noiseless
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Noisy
matching $O(\log N)$ random points of a noisy sinusoid
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ \hline 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ 1 & W^{3} & W^{6} & \cdots & W^{3 \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{k} & W^{2 k} & \cdots & W^{k \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \ldots & W^{19 \times 19}\end{array}\right]$

## Noisy Setting: R-FFAST | Sample-Optimal Recovery with Near Linear Time

Theorem: R-FFAST algorithm computes the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$,

- using $M$ noisy samples, where $M=O(K \log N)$, (order optimal)
- with probability at least $1-O(1 / M)$,
- in $O(N \log N)$ computations.
- for a finite SNR.

Theorem: FFAST algorithm computes the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$,

- using $M=O(K)$ samples,
- in $O(K \log K)$ computations,
- with probability at least $1-O(1 / M)$.


## Noisy Setting: R-FFAST | Sample-Optimal Recovery with Near Linear Time

- $K=20$ and 40
- $N=0.124$ million
- Sample cost $M=2940$



## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time

Theorem: R-FFAST algorithm computes the $K$-sparse DFT of $\boldsymbol{x} \in \mathbb{C}^{N}$,

- using $M$ noisy samples, where $M=O\left(K \log ^{1.33} N\right)$,
- with probability at least $1-O(1 / M)$,
- in $O\left(K \log ^{2.33} N\right)$ computations,
- for a finite and sufficiently high SNR.


## Noiseless

matching 2 points of a noiseless sinusoid
Noisy
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \cdots & W^{19 \times 19}\end{array}\right]$
matching $O(\log N)$ random points of a noisy sinusoid
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ \hline 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ 1 & W^{3} & W^{6} & \cdots & W^{3 \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{k} & W^{2 k} & \cdots & W^{k \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \cdots & W^{19 \times 19}\end{array}\right]$

## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time

## Noisy

## Noisy (sub-linear)

matching $O(\log N)$ random points of a noisy sinusoid
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ \hline 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ 1 & W^{3} & W^{6} & \cdots & W^{3 \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{k} & W^{2 k} & \cdots & W^{k \times 19} \\ \hline \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \ldots & W^{19 \times 19}\end{array}\right]$
$\left[\begin{array}{ccccc}1 & 1 & 1 & \cdots & 1 \\ 1 & W & W^{2} & \cdots & W^{19} \\ \hline 1 & W^{2} & W^{4} & \cdots & W^{2 \times 19} \\ 1 & W^{3} & W^{6} & \cdots & W^{3 \times 19} \\ \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{k} & W^{2 k} & \cdots & W^{k \times 19} \\ \vdots & \ddots & \ddots & \ddots & \vdots \\ 1 & W^{19} & W^{19 \times 2} & \cdots & W^{19 \times 19}\end{array}\right]$
each piece is of length $O\left(\log ^{0 . \dot{3}} \mathrm{~N}\right)$

## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time



## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time



1234

- For each random start, take consecutive delays spaced by $2^{i}$



Noisy

$\mathrm{O}(\log \mathrm{N})$ random starts

## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time



- For each random start, take consecutive delays spaced by $2^{i}$
- [1989'Kay] provides an unbiased and efficient estimate of $2^{i} \omega$



## Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time



- For each random start, take consecutive delays spaced by $2^{i}$
- [1989'Kay] provides an unbiased and efficient estimate of $2^{i} \omega$
- $O\left(\log ^{1 / 3} N\right)$ consecutive rows are sufficient for matching each piece

Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time


- Piece 1: estimates $\omega$ with no ambiguity

Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time


- Piece 1: estimates $\omega$ with no ambiguity
- Piece 2: estimates $2 \omega \Longrightarrow$ unwrapping leads to 2 ambiguities

Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time


- Piece 1: estimates $\omega$ with no ambiguity
- Piece 2: estimates $2 \omega \Longrightarrow$ unwrapping leads to 2 ambiguities
- Piece $i$ : estimates $2^{i} \omega \Longrightarrow$ unwrapping leads to $2^{i}$ ambiguities

Noisy Setting: R-FFAST | Near Sample-Optimal Recovery with Sub-linear Time


- accuracy improves dyadically $2^{i}$
- each piece provides 1 information bit
- $\log N$ pieces $\Longrightarrow N$ locations


## Computing Sparse DFT | Some Concrete Examples of FFAST



Original Brain image


Fourier domain of the Brain Image


Brain image reconstructed by FFAST

## Computing Sparse DFT | Numerical Comparisons




Sub-linear time performance:

- Signal length increased 15 fold.
- Processing time less than 3 fold.


## Sample Complexity:

- This verifies the sample overhead of fast search over the slow search


## Discussions:

- Sparse WHT
- Compressed Sensing


## Walsh-Hadamard Transform | What is it?

IEEE TRANSACTIONS ON SIGNAL PROCESSING, VOL. 41, NO. 3, MARCH 1993

## The Poorman's Transform: Approximating the Fourier Transform without Multiplication

Michael P. Lamoureux


## Walsh-Hadamard Transform |

- Noiseless results [2013'Scheibler et al] Scheibler, R., Haghighatshoar, S., Vetterli, M., 2013 51st Annual Allerton Conf. on Communication, Control, and Computing (pp. 1250-1257).


## Walsh-Hadamard Transform

- Noiseless results [2013'Scheibler et al] Scheibler, R., Haghighatshoar, S., Vetterli, M., 2013 51st Annual Allerton Conf. on Communication, Control, and Computing (pp. 1250-1257).
- Noisy results [2014'Li et al] Xiao Li; Bradley, J.K.; Pawar, S.; Ramchandran, K., "The SPRIGHT algorithm for robust sparse Hadamard Transforms," Information Theory (ISIT), 2014 IEEE International Symposium on , vol., no., pp.1857,1861, June 29 2014-July 42014.


Sparse Iterative Graph-based Hadamard Transform (SPRIGHT)

## Conclusion

- FFAST algorithm for computing $K$-sparse DFTs


## C + + code available!

- exploits coding-theory principle (i.e., sparse-graph alias codes)
- Noiseless:
* $M=O(K)$ samples and $T=O(K \log K)$ run-time
- Noisy:
* sample-optimal recovery $M=O(K \log N)$ with near-linear run-time $T=O(N \log N)$
* near sample-optimal recovery $M=O\left(K \log ^{1.3} N\right)$ with sub-linear run-time $T=O\left(K \log ^{2.3} N\right)$
- Extensions to sparse WHT and compressed sensing
- Sparse WHT
https://www.eecs.berkeley.edu/~ kannanr/assets/project_ffft/WHT_noisy.pdf
- Compressed sensing using sparse-graph codes http://www.eecs.berkeley.edu/~xiaoli/FR_CS_SGC.pdf


## Future Directions

- More general sparsity and signal models
- Off-grid frequency estimation
- Practical applications in
- MRI
- Optical imaging such as Fourier Ptychography
- Phase retrieval $\Longrightarrow$ "PhaseCode" design http://arxiv.org/abs/1408.0034

